

NATIONAL EXAMS May 2010  
07-Elec-B2 Advanced Control Systems

3 hours duration

NOTES:

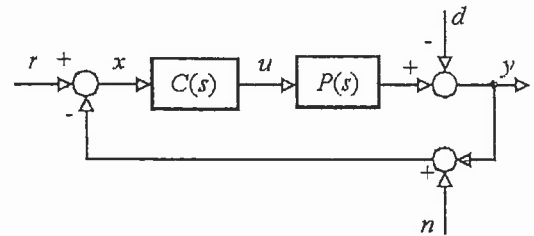
1. If a doubt exists as to the interpretation of any question, the candidate is urged to submit with the answer paper, a clear statement of any assumptions made.
2. Candidates may use one of two calculators, a Casio or Sharp approved models.
3. This is a closed-book examination. Tables of Laplace and z-transforms are attached as page 4 and page 5.
4. Any four questions constitute a complete paper. Only the first four questions as they appear in your answer book will be marked.
5. All questions are of equal value.

07-Elec-B2 Advanced Control Systems – May 2010

1. Consider the automobile cruise control system below with,

$$C(s) = \frac{5}{20s + 1}, \quad P(s) = \frac{1}{(4s + 1)^2}$$

- (a) Determine the steady state error,  $e = r - y$ , when the grade is level, i.e.,  $d = 0$ , the measurement noise,  $n = 0$ , and the setpoint speed is  $r = 8$ .
- (b) Suddenly the grade increases such that  $d = 3$ , (with  $r = 8$ ). Determine the new steady state error.
- (c) Now,  $r(t) = 8 \sin(0.03t)$ . Determine the steady state error,  $e$ .
- (d) Re-design the controller,  $C(s)$ , such that the steady state error,  $e$ , is zero when  $r(t)$  and  $d(t)$  are arbitrary constants, measurement noise,  $n = 0$ , and the closed loop system is stable with a gain margin of at least 6dB.



2. Consider the system,  $P(s) = \frac{16(\alpha s + 1)}{s(s + 2)^2}$ .

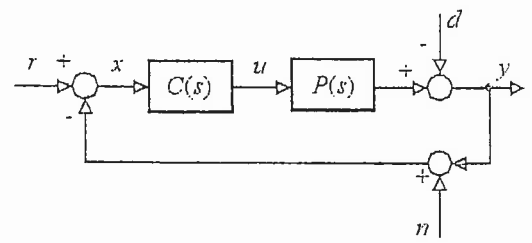
- (a) Find a state space model for the system.
- (b) Justify the conditions under which the system controllable? observable?
- (c) The system input and output are uniformly sampled with a sample period of  $h$  and the discrete input is applied to a zero order hold device. Determine the poles of the sampled data system as a function of  $h$ . Detailed calculations are not necessary.

3. Several experiments are conducted on an unknown plant:

When a step of magnitude 2 is applied to the input, the steady state output is 10.  
 When a sinusoid of frequency 8 rad/sec is applied, the phase lag at the output is  $90^\circ$ .  
 When a sinusoid of frequency 5 rad/sec is applied, the phase lag at the output is  $15^\circ$ .

- (a) Assume the system,  $P(s)$ , is second order system and has no finite zeros. Find the parameters of the second order model.
- (b) For the model identified in (a) determine the maximum overshoot for a unit step input.
- (c) Sketch the polar (Nyquist) plot for the identified model being careful to label axes, scales, and intercepts.
- (d) Justify whether the system is stable or not when a controller,  $C(s) = 1/s$ , is cascaded with  $P(s)$  in a negative (unity) feedback loop.

4. Consider the feedback system below with,  $P(s) = \frac{4}{s^2 + 9}$ .

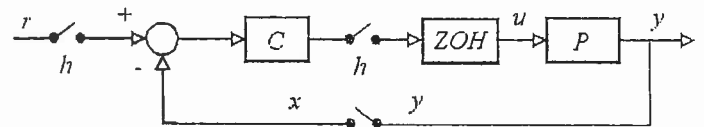


(a) Determine a proper and stable  $C(s)$  such that the transfer function that relates  $d$  to  $y$  is:  $\frac{b_3s^3 + b_2s^2 + b_1s + b_0}{(s + 2)^2(s + 1)}$ . Recall

that  $C(s)$  is proper if the degree of the numerator is less than or equal to that of the denominator.

(b) Determine the transfer function that relates  $r$  to  $y$  and the transfer function that relates  $n$  to  $x$ .

5. Consider the sampled data system shown on the right. The input to the ZOH, the set-point,  $r$ , and the output,  $y$ , are uniformly sampled with a sample period of  $h = 0.1$  with  $C(z)$  and  $P(s)$  given by,



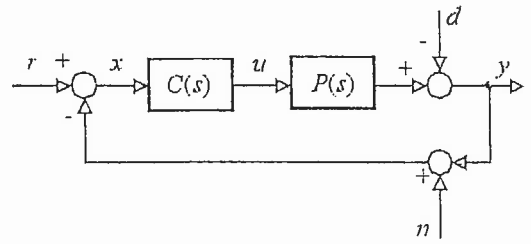
$$C(z) = K, \quad P(s) = \frac{e^{-sh}}{s}$$

- (a) Determine the discrete closed loop transfer function,  $T(z)$ , that relates  $X(z)$  to  $R(z)$ .
- (b) Determine the limiting value of  $K$  for stability.
- (c) Consider replacing  $C(z)$  by a statefeedback controller. Design the statefeedback controller (assuming states  $y(k)$  and  $y(k + 1)$  are available for feedback) such that the closed loop poles are all at  $z = 0.5$

6. Consider the (continuous time) feedback system below with,  $C(s) = K$ ,  $P(s) = \frac{2e^{-0.4s}}{s + 1}$ .

- (a) Determine the range of  $K$  that results in closed loop stability.
- (b) Determine the phase margin when  $K = 1$  and sketch the associated Nyquist plot.
- (c) The system is stable and operating with noise,  $n(t) = 0$ , disturbance,  $d(t) = 1$ , and set-point,  $r(t) = 0$ . Determine the tracking error,  $e(t) = r(t) - y(t)$ , as a function of  $K$ .

(d) It is necessary that the steady state error is zero for any arbitrary but constant value of  $d$ . Choose a new controller,  $C(s)$ , that ensures the steady state tracking error is zero *and* the closed loop system is stable with a gain margin of at least 4db.



Inverse Laplace Transforms	
$F(s)$	$f(t)$
$\frac{A}{s + \alpha}$	$Ae^{-\alpha t}$
$\frac{C + jD}{s + \alpha + j\beta} + \frac{C - jD}{s + \alpha - j\beta}$	$2e^{-\alpha t} (C \cos \beta t + D \sin \beta t)$
$\frac{A}{(s + \alpha)^{n+1}}$	$\frac{At^n e^{-\alpha t}}{n!}$
$\frac{C + jD}{(s + \alpha + j\beta)^{n+1}} + \frac{C - jD}{(s + \alpha - j\beta)^{n+1}}$	$\frac{2t^n e^{-\alpha t}}{n!} (C \cos \beta t + D \sin \beta t)$

Inverse z-Transforms	
$F(z)$	$f(nT)$
$\frac{Kz}{z - a}$	$Ka^n$
$\frac{(C + jD)z}{z - re^{j\varphi}} + \frac{(C - jD)z}{z - re^{-j\varphi}}$	$2r^n (C \cos n\varphi - D \sin n\varphi)$
$\frac{Kz}{(z - a)^r}, r = 2, 3, \dots$	$\frac{Kn(n-1)\dots(n-r+2)}{(r-1)!a^{r-1}} a^n$

Table of Laplace and z-Transforms ( $h$ denotes the sample period)		
$f(t)$	$F(s)$	$F(z)$
unit impulse	1	1
unit step	$\frac{1}{s}$	$\frac{z}{z-1}$
$e^{-\alpha t}$	$\frac{1}{s+\alpha}$	$\frac{z}{z-e^{-\alpha h}}$
$t$	$\frac{1}{s^2}$	$\frac{hz}{(z-1)^2}$
$\cos \beta t$	$\frac{s}{s^2+\beta^2}$	$\frac{z(z-\cos \beta h)}{z^2-2z\cos \beta h+1}$
$\sin \beta t$	$\frac{\beta}{s^2+\beta^2}$	$\frac{z \sin \beta h}{z^2-2z\cos \beta h+1}$
$e^{-\alpha t} \cos \beta t$	$\frac{s+\alpha}{(s+\alpha)^2+\beta^2}$	$\frac{z(z-e^{-\alpha h} \cos \beta h)}{z^2-2ze^{-\alpha h} \cos \beta h+e^{-2\alpha h}}$
$e^{-\alpha t} \sin \beta t$	$\frac{\beta}{(s+\alpha)^2+\beta^2}$	$\frac{ze^{-\alpha h} \sin \beta h}{z^2-2ze^{-\alpha h} \cos \beta h+e^{-2\alpha h}}$
$t f(t)$	$-\frac{dF(s)}{ds}$	$-zh \frac{dF(z)}{dz}$
$e^{-\alpha t} f(t)$	$F(s+\alpha)$	$F(ze^{\alpha h})$